



APPLICATION NOTE 4215

Estimating Switching Speed for the DS2712 NiMH Battery Charger

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Abstract: In switch mode, the switching speed of the DS2712 battery charger varies based on inductor selection, charge current, and battery voltage. In some cases, the speed under various conditions should be known to account for noise sensitivity, peak-to-peak charge current, and inductor selection constraints. This application note provides an easy-to-use spreadsheet for estimating a circuit's switching speed with different component values and battery parameters.

Introduction

When designing with the [DS2712](#) switch-mode battery charger, it is useful to quickly determine the range of frequencies within which the circuit can be expected to transition. Knowing and controlling the switching frequency can help optimize inductor selection, give guidelines for transistor selection, and dictate PCB layout considerations, among other things. For the DS2712 architecture, the charge mode does not affect switching frequency (as it does with the [DS2715](#) platform) because the reference voltage is fixed in all modes. While the inductor plays the largest role in setting the switching frequency, the variation in battery cell voltages is the dominant variable effect. This application note presents an analysis of the DS2712 charger circuit, along with an easy-to-use spreadsheet for estimating the switching frequency.

Circuit Overview

Figure 1 shows the equivalent circuit for a buck switching charger with a battery and sense resistor load. The DS2712 architecture operates by regulating the voltage across R_{SENSE} to a preset value for all charging modes. The sense resistor value, along with this preset reference voltage, then controls the value of I_{CHARGE} . For NiMH batteries, four stages of charging are preferred—all of which (Precharge, Fast-Charge, Topoff, and Maintenance) are regulated through the voltage across the sense resistor in the DS2712. The precharge and topoff states operate at 25% of the average current regulated during fast-charge. Maintenance mode operates at 1/64th the fast-charge current.

The various charge states are entered depending on the state of the battery stack. When the battery voltage is less than 1.0V, precharge mode is used to slowly charge depleted batteries. Fast-charge mode is only used when the battery voltage is above 1.0V and not more than 1.65V, but may also terminate when other fast-charge conditions are met. Topoff mode follows fast-charge mode for a preset period of time. $R_{PARASITIC}$ is an optional parameter that is added to the reference circuit to represent any significant parasitic voltage drops that may occur as a result of battery, fuse, or other connection nonidealities.

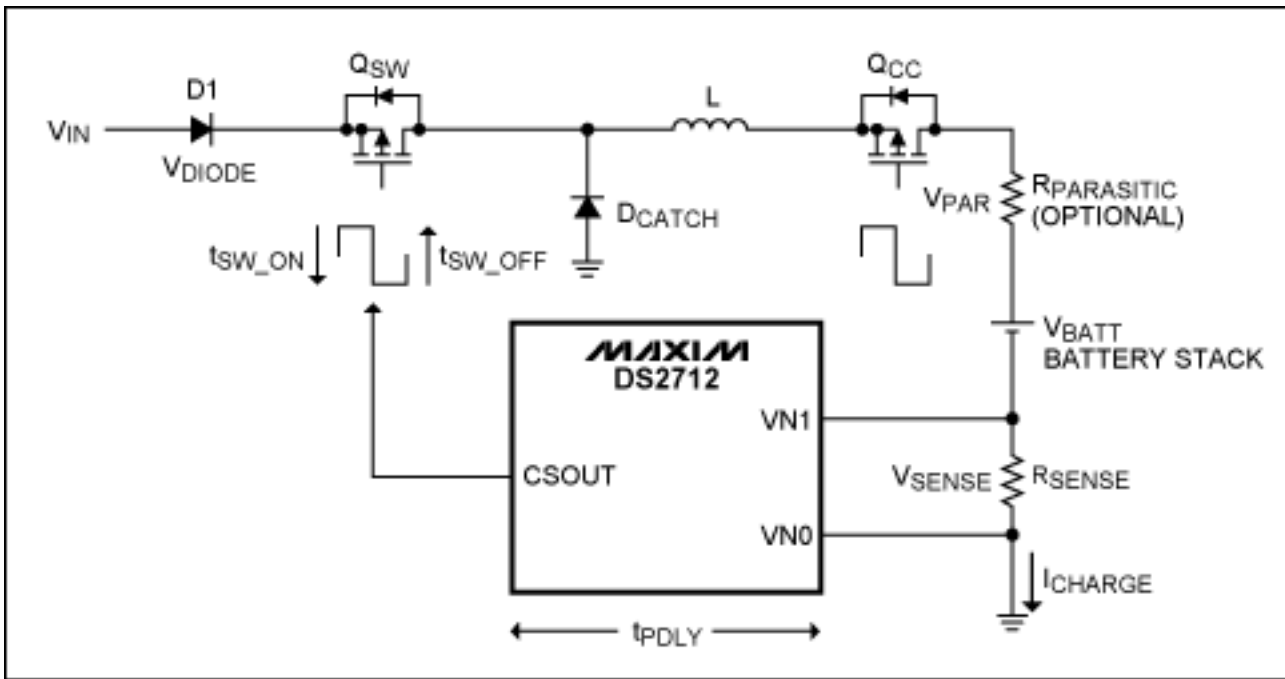


Figure 1. The DS2712 switching reference circuit.

The control circuit for regulating the sense resistor voltage consists of a comparator with hysteresis and a reference voltage, which drives an open-drain output pin. Changes to the duty cycle for the various switching modes is handled by Q_{CC} and the CC output pin (not shown). Switching speed while Q_{CC} is conducting is determined by the amount of time needed for I_{CHARGE} to go from one peak to the other and back again, which includes any delays in the feedback circuit and in the process of turning Q_{SW} and D_{CATCH} off and on. These delays cause I_{CHARGE} to overshoot one or both of the comparator thresholds, creating a larger peak-to-peak variation than nominal. **Figure 2** shows an example of ideal switching control waveforms, while **Figure 3** shows waveforms with extra turn-off delays in Q_{SW} .

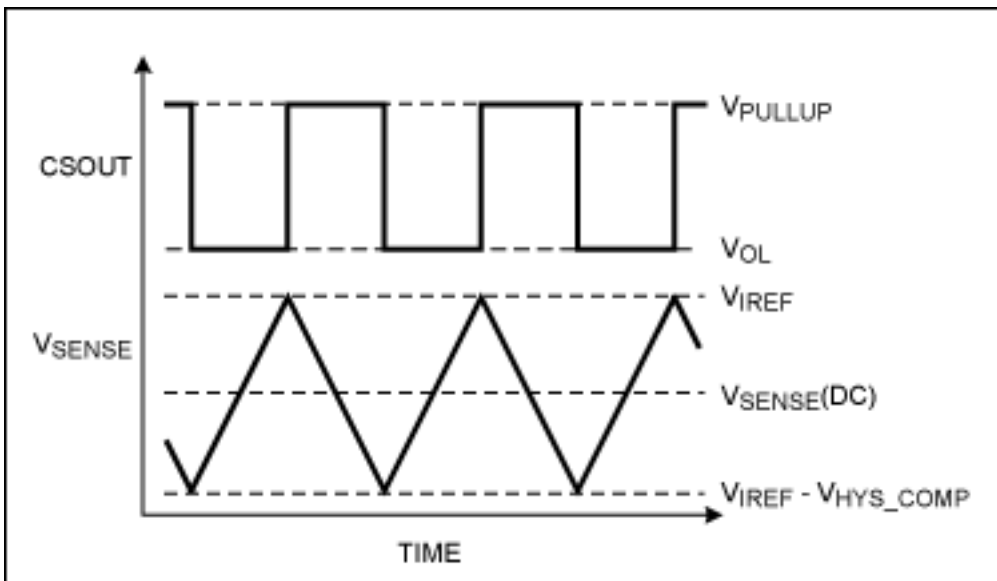


Figure 2. Ideal charge-current control waveforms. (Not drawn to scale.)

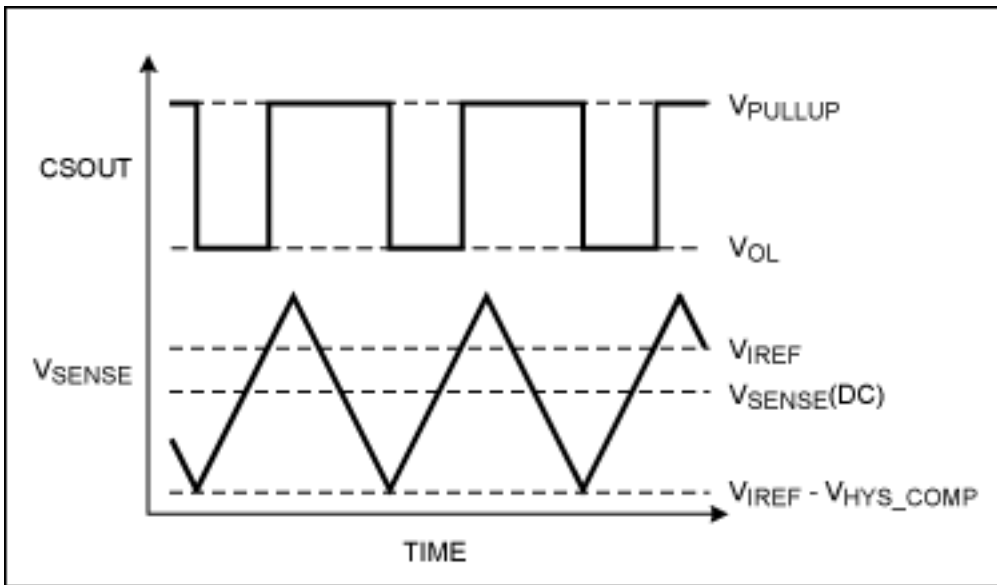


Figure 3. Charge-current control waveforms with slow Q_{SW} turn-off. (Not drawn to scale.)

Calculation Method

The rate of change for I_{CHARGE} is governed by the voltage and current relationship of an ideal inductor L .

$$V = L \times di/dt \text{ and } dt = L \times di/V \quad (\text{Eq. 1})$$

Hence, the amount of time for the on and off states of Q_{SW} varies with the voltage developed across L in each state, and the change in I_{CHARGE} from peak to peak. The peak-to-peak value is set by the hysteresis value of the charge mode and any control circuit delays that cause overshoot of the comparator levels. See the data sheet for the DS2712 device for definitions of the sense comparator reference voltage (V_{IREF}) and the hysteresis voltage (V_{HYS_COMP}).

In order to simplify the switching-speed approximation, let us make the assumption that the voltage across the inductor for each Q_{SW} state stays constant. Let us also assume that the midpoint of V_{SENSE} from peak to peak [$V_{IREF} - (0.5 \times V_{HYS_COMP})$] provides a reasonable steady-state condition for calculating the inductor voltage of each state. With those simplifications in place, the voltage drop around the loop of the reference circuit in Figure 1 for the Q_{SW_ON} and Q_{SW_OFF} states can be calculated to yield the inductor voltage for a given battery voltage as follows:

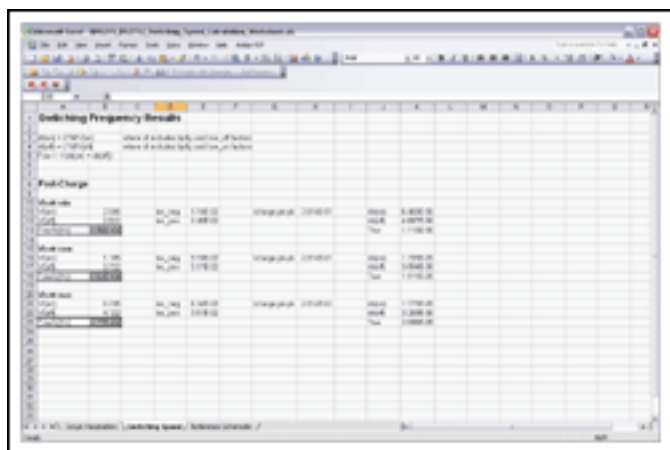
$$V_L(Q_{SW_ON}) = V_{IN} - V_{DIODE} - V_{QSW} - V_{QCC} - V_{PAR} - V_{BATT} - V_{SENSE} \quad (\text{Eq. 2})$$

$$V_L(Q_{SW_OFF}) = V_{QCC} + V_{PAR} + V_{BATT} + V_{SENSE} + V_{DCATCH} \quad (\text{Eq. 3})$$

Since the battery voltage may vary over a 1.0V to 1.65V range in fast-charge mode, it is useful to calculate the inductor voltage and then the switching speed at several operating points. A spreadsheet is available for [download](#) that will perform the calculations described in this application note for three different battery voltages in fast-charge mode, and at one battery stack voltage in precharge and toff modes.

Once the inductor voltages for each state and each operating point are calculated, any delays in the control circuit may be factored into the on and off times for Q_{SW} . For our approximation, this is accomplished by first calculating the extra change in I_{CHARGE} due to the turn-on and turn-off delays of Q_{SW} according to the following

switching-speed result. Included among these are the inductor voltages for each state, the change in I_{CHARGE} due to delays, and the individual on and off state times.



The image shows a screenshot of a spreadsheet application with a window titled 'Switching Frequency Results'. The spreadsheet contains several columns and rows of data. The first column lists parameters such as 'Inductor Voltage', 'Change in I_CHARGE', and 'On/Off State Times'. The subsequent columns contain numerical values for these parameters. The spreadsheet is organized into sections, with some rows highlighted in grey. The overall layout is typical of a data analysis tool used for engineering calculations.

[More detailed image](#) (PDF, 113kB)

Figure 5. The switching-speed results tab.

Conclusions

Many useful pieces of information can be taken from the estimations of switching speed as they relate to the reference circuit. It can be seen that the main factors affecting the switching speed of the DS2712 switch-mode NiMH charger are the inductor value, I_{CHARGE} , the headroom of V_{IN} above the battery stack voltage, and in some cases delays resulting from the slow switching of Q_{SW} . The analysis presented here can be used to estimate switching speed for these circuits and develop an understanding of how best to adjust this speed when necessary.

Application Note 4215: www.maxim-ic.com/an4215

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